

► First, do a summary of things learned so far!!!

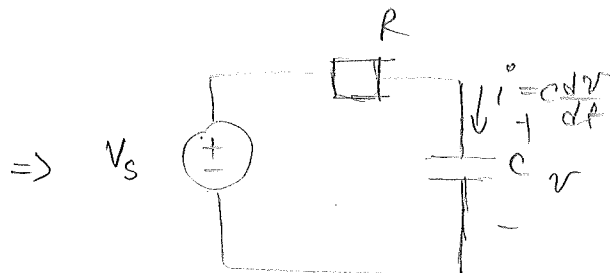
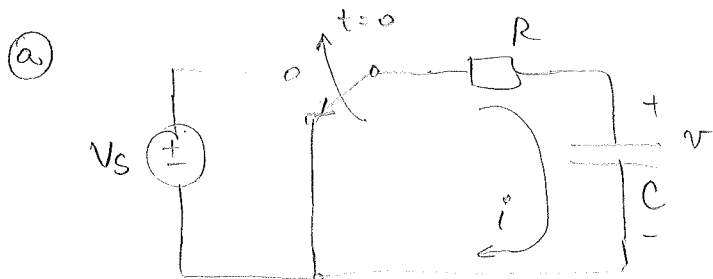
Transient response of first-order circuits

- We had found that the complete response to a dc forcing function consists of a transient component & a dc steady-state component.
- The transient response has the same functional form as the natural response, which is an exponentially decaying function!
- The dc steady-state component has the same functional form as the forcing function, which is a constant.

Comparison of C and L behavior:

Element:	C	L
• dc steady-state condition	$v_C = \text{const}$	$i_L = \text{const}$
• dc steady-state behavior	open circuit	short circuit
• <u>cannot</u> change instantaneously	$v_C$	$i_L$
• can change instantaneously	$i_C$	$v_L$

1 The RC circuit



capacitance is initially discharged!  
 $i(0^-) = 0$     $v(0^-) = 0$ .

Recall: (1)  $RC \cdot \frac{dv}{dt} + v = V_s$

which begins at  $t = 0^+$ !

and property (or "crucial observation"):  $v(0^-) = v(0^+) = 0 \text{ V}$

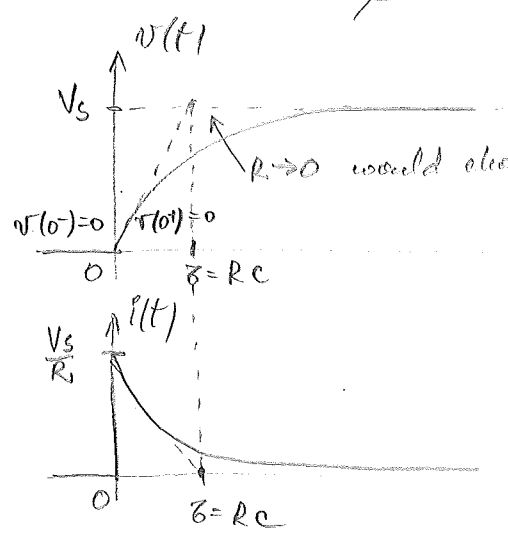
Also recall:  $y(t) = y(0^-) \cdot e^{-\frac{t}{\tau}} + X_s (1 - e^{-\frac{t}{\tau}})$   
 $\tau = RC$   
 $y(0^-) = 0$   $X_s = V_s$

So, for  $t \geq 0$ , the solution is:

$v(0^-) = 0; t \leq 0$   $v(t) = V_s (1 - e^{-\frac{t}{RC}})$  (2);  $t \geq 0$  !

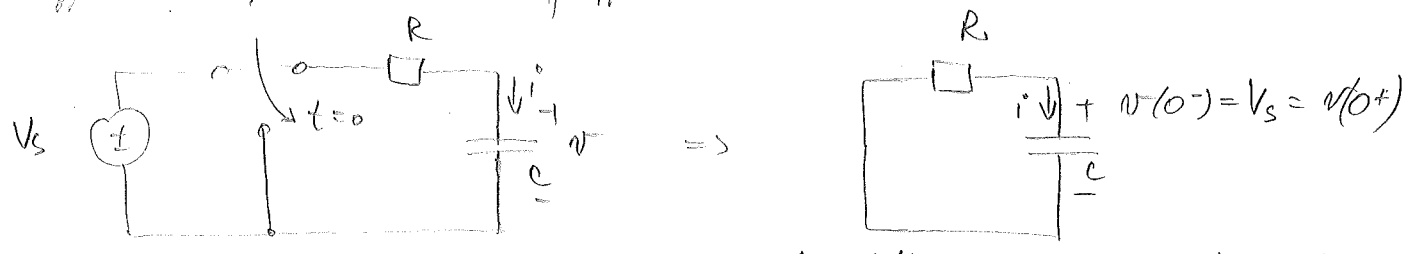
← basically just the forced component!

Also:  $i = C \frac{dv}{dt} = C \cdot V_s \cdot \frac{1}{RC} e^{-\frac{t}{RC}} = \frac{V_s}{R} e^{-\frac{t}{RC}} = i(t)$  (3);  $t \geq 0$  !  
 $i(0^-) = 0; t \leq 0$



$R \rightarrow 0$  would change the capacitance "instantaneously" but this would require an  $\infty$  current spike!  
 : Referred to as a "spike"

(b) Suppose now, the switch is flipped back!

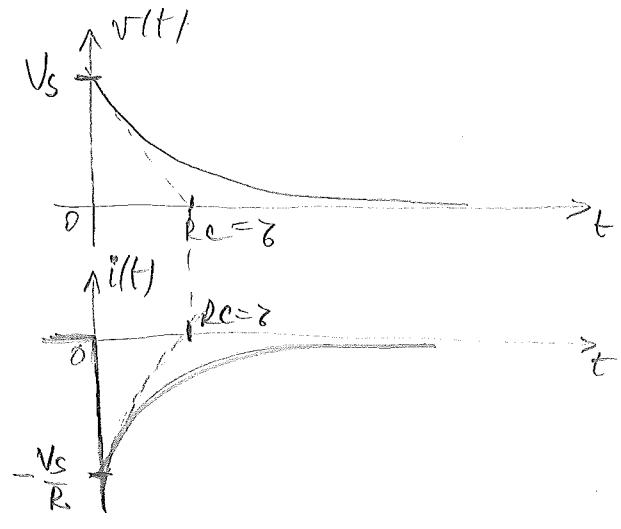


- And lets choose again  $t=0$  this instant for convenience.
- By the continuity rule for capacitance:  
 $v(0^-) = v(0^+) = V_s$
- This is the over-free circuit! with the solution:

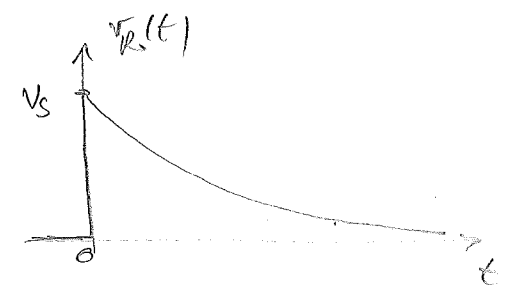
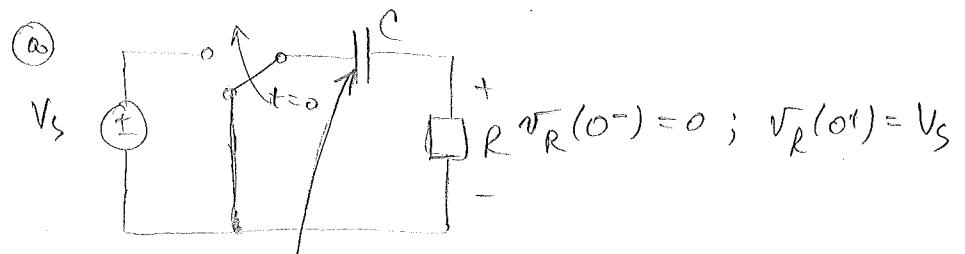
(4)  $RC \frac{dv}{dt} + v = 0$

(5)  $v(t) = V_s \cdot e^{-\frac{t}{RC}}$ ;  $t \geq 0$   
 $v(0^-) = V_s$

Also:  $i = C \frac{dv}{dt} = -C \cdot V_s \cdot \frac{1}{RC} e^{-\frac{t}{RC}} = -\frac{V_s}{R} e^{-\frac{t}{RC}} = i(t)$  (6)  $t \geq 0$   
 $i(0^-) = 0; t \leq 0$

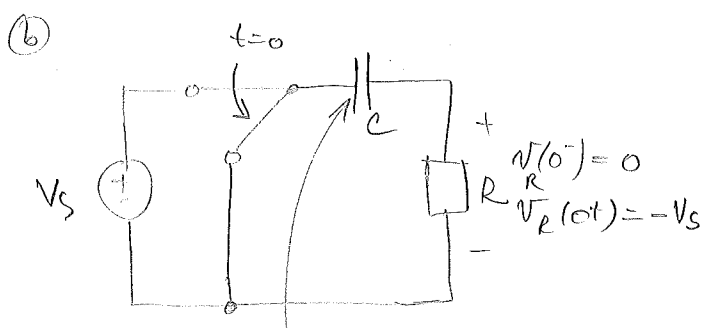


1 The RC circuit To-1 to prepare a de la Broopra or give it as the assignment



- At  $t=0$  left plate jumps from  $0V \rightarrow V_S$
- By the continuity rule right plate also jumps to  $V_S$ !

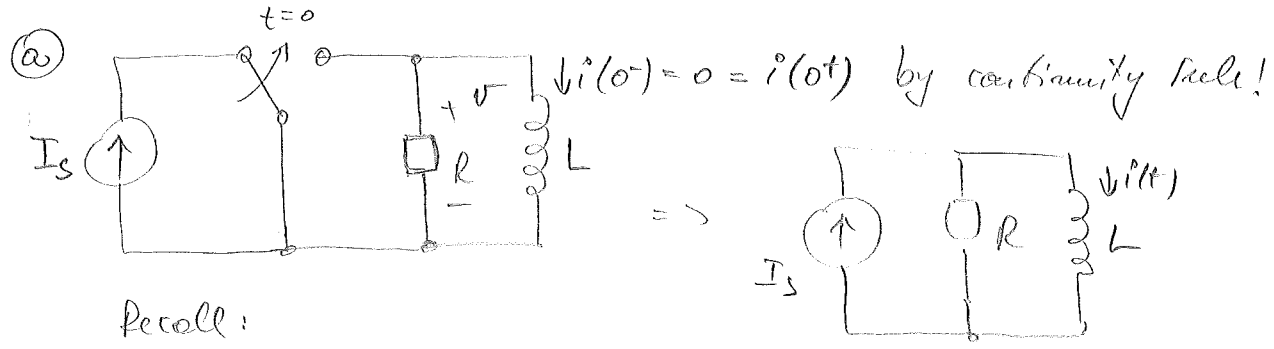
From equation (3):  $i(t) = \frac{V_S}{R} e^{-\frac{t}{RC}} \Rightarrow v_R(t) = V_S \cdot e^{-\frac{t}{RC}} \quad t \geq 0$



Jumps  $V_S \rightarrow 0$ ; right plate jumps from  $0 \rightarrow -V_S$  due to the continuity rule! So,  $v_R(0^+) = -V_S$ ; R draws current and discharges.

By equation (6):  $i(t) = -\frac{V_S}{R} e^{-\frac{t}{RC}} \Rightarrow v_R(t) = -V_S e^{-\frac{t}{RC}} \quad t \geq 0.$

**2) RL circuit**



(1)  $\frac{L}{R} \frac{di}{dt} + i = I_s$

Also recall:  $y(t) = y(0) \cdot e^{-\frac{t}{\tau}} + X_s (1 - e^{-\frac{t}{\tau}})$ ;  $\tau = \frac{L}{R}$

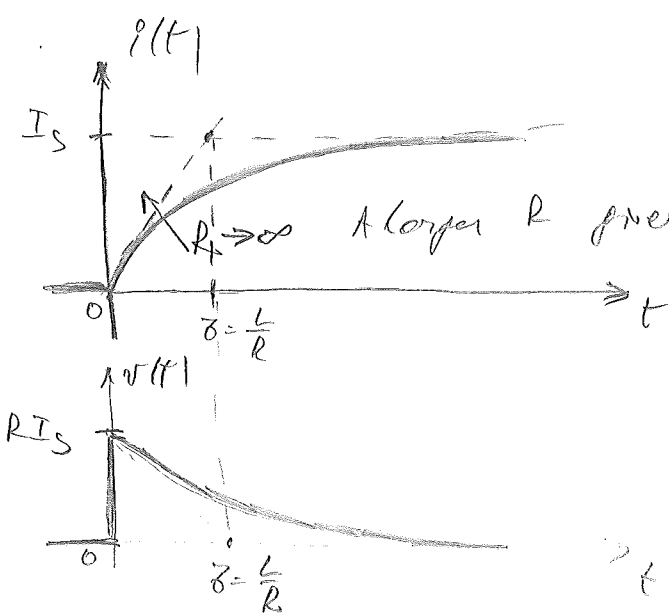
Annotations:  $y(t) \rightarrow i(t)$ ,  $y(0) \rightarrow i(0) = 0$ ,  $X_s \rightarrow I_s$

Hence  $i(t) = I_s (1 - e^{-\frac{R}{L}t})$  (2)  $t \geq 0$

Condition:  $i(0) = 0, t \leq 0$

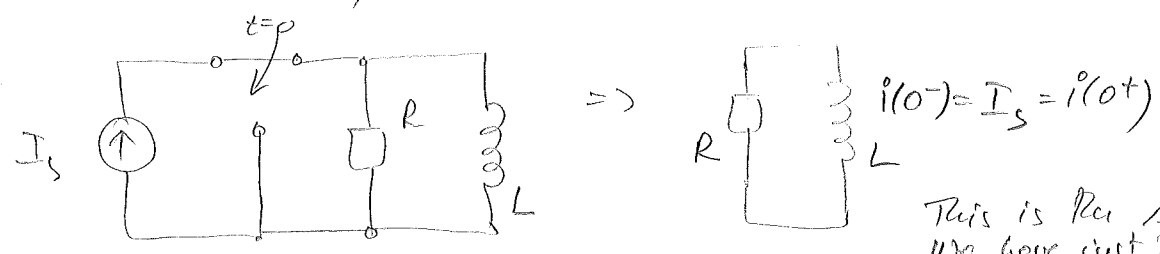
Also:  $v = L \frac{di}{dt} = -L \cdot (-\frac{R}{L}) \cdot I_s \cdot e^{-\frac{R}{L}t} = R I_s \cdot e^{-\frac{R}{L}t} = v(t)$  (3)  $t \geq 0$

Condition:  $v(0) = 0, t \leq 0$



A larger  $R$  gives a larger  $RI_s$  a voltage, which gives a faster rate of current buildup!

(b) Leave (a) for very long time, then flip back the switch!



This is the source-free circuit!  
We have just the natural response!

$$(4) \left[ \frac{L}{R} \frac{di}{dt} + i = 0 \right]$$

$$(5) \left[ i(t) = I_s \cdot e^{-\frac{R}{L}t} \right] \quad t \geq 0.$$

$$i(0^-) = I_s$$

$$\text{Also: } v = L \frac{di}{dt} = -I_s \cdot \frac{R}{L} \cdot e^{-\frac{R}{L}t} = -RI_s \cdot e^{-\frac{R}{L}t} = v(t) \quad (6) \quad \underline{t \geq 0}$$

$$v(0^-) = 0$$

