

Examples

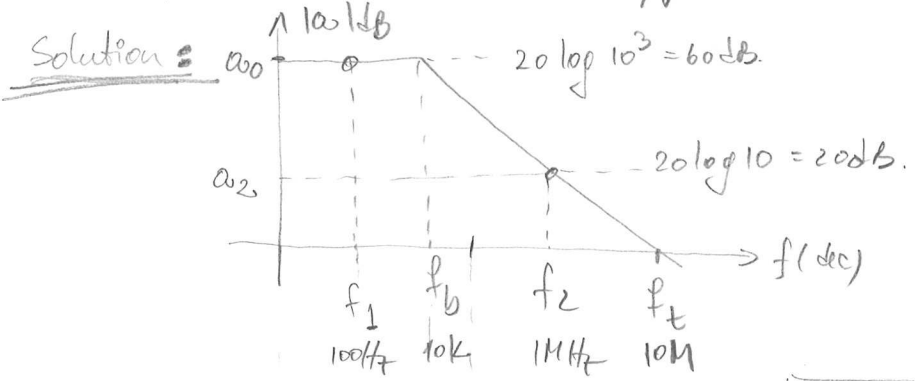
6.3. (Problem 6.3 from textbook, page 303)
 Constant GBP opamp has
 of freq. less than f_b .

$$|a(j100\text{Hz})| = 1\text{V/mV} = 1000\text{V/V}$$

$$|a(j1\text{MHz})| = 10\text{V/V}$$

Example 1:
 (data to ex. 1. in 2011)

- a) $f_x = ?$ at which $\angle a = -60^\circ = \angle a(jf_x)$
- b) $f_x = ?$ at which $|a| = 2\text{V/V}$



$$GBP = f_t = a_f \cdot f = a_2 \cdot f_2 = 10 \cdot 1\text{MHz} = \boxed{10\text{MHz} = f_t}$$

$$\text{Also: } f_b = \frac{f_t}{a_0} = \frac{10\text{MHz}}{10^3} = \boxed{10\text{kHz} = f_b}$$

a) $a(jf) = \frac{a_0}{1 + jf/f_b} \Rightarrow \angle a = \angle e(jf) = -\arctan(f/f_b) = -\tan^{-1}(f/f_b)$

$$-60 = -\tan^{-1}\left(\frac{f_x}{f_b}\right)$$

$$60 = \tan^{-1}\left(\frac{f_x}{10^4}\right) \Rightarrow \boxed{f_x = 17.32\text{kHz}}$$

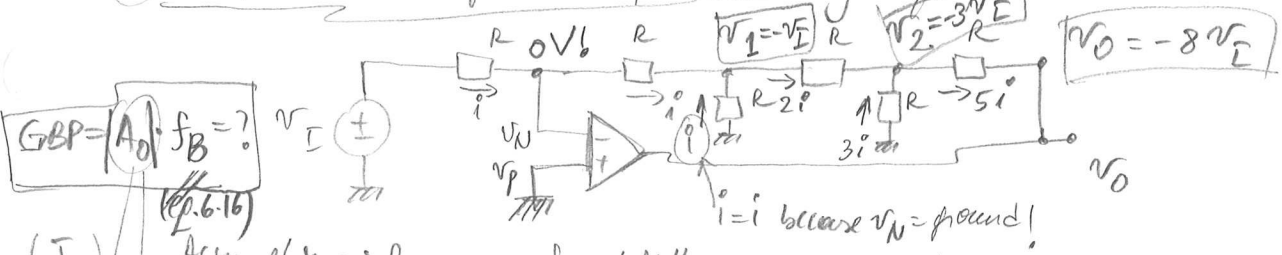
expected!

b) $|a(jf)| = \frac{a_0}{\sqrt{1 + (f/f_b)^2}} \Big|_{f=f_x} = 2\text{V/V} \Rightarrow \left(\frac{a_0}{2}\right)^2 = 1 + \left(\frac{f_x}{f_b}\right)^2$

$$\frac{10^6}{4} = 1 + \frac{f_x^2}{(10^4)^2} \Rightarrow \boxed{f_x = 5\text{MHz}}$$

6.80 Example 2

(a) Find the closed-loop GBP of the inverting amplifier:



GBP = $A_0 \cdot f_B = ?$
 (Eq. 6.16)

(I) $A_0 = ? = A_{ideal}$

Assume $r_d \gg R$
 $r_o \ll R$

$f_t = 4 \text{ MHz}$. ← Given!

$i = \frac{v_I}{R}$

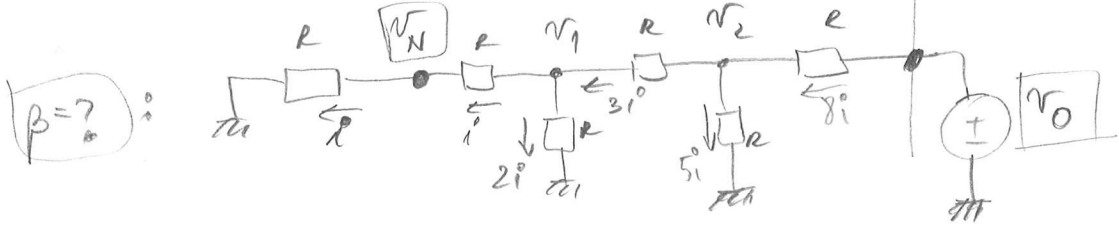
$v_1 = -v_I$, $v_2 = v_1 - R \cdot 2i = -3v_I$, $v_O = v_2 - R \cdot 5i = -8v_I$

$\Rightarrow A_{ideal} = -8 \frac{v_O}{v_I} = A_0$

(II) $f_B = ?$

$f_B = \beta \cdot f_t$ (see Eq. 6.15(b)).
 $\Rightarrow f_B = 4 \text{ MHz}$

Negative feedback taken separately to study $\beta = \frac{v_N}{v_O}$



$\beta = ?$

we disconn. v_N to acc. to null point to null point to IV amplifier.
~~neg. feedback network.~~
~~short out \Rightarrow source voltage~~
~~short in \Rightarrow output current~~
 Do NOT mention if confusion is! This is used as voltage amplifier!

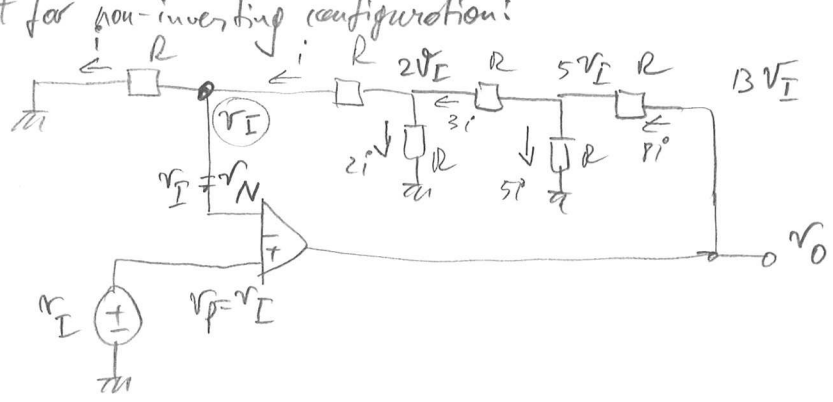
$i = \frac{v_N}{R}$, $v_1 = 2v_N$; $v_2 = v_1 + R \cdot 3i = 5v_N$; $v_O = v_2 + 8R \cdot i = 13v_N$

$\Rightarrow \beta = \frac{v_N}{v_O} = \frac{v_N}{13v_N} = \frac{1}{13} \frac{V}{V}$

$\Rightarrow f_B = \frac{4 \text{ MHz}}{13} \cdot \frac{V}{V} \Rightarrow \boxed{\text{GBP} = A_0 \cdot f_B = 8 \cdot \frac{4 \text{ MHz}}{13} = \frac{32}{13} \text{ MHz}}$

(b)

Repeat for non-inverting configuration:



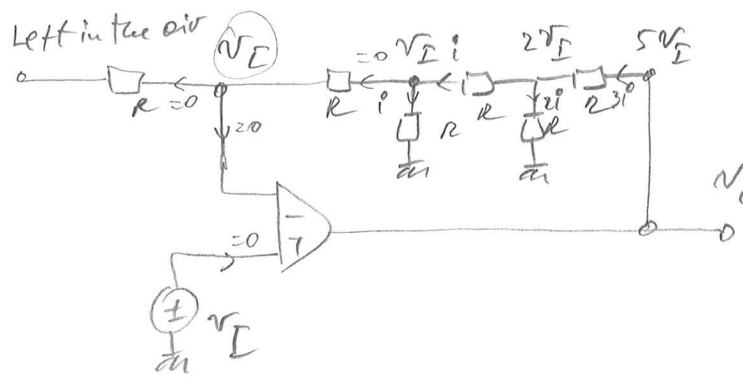
(3) : same current ratios as in part (a) from (a).

Left as exercise!

$$\Rightarrow A_0 = 13 \frac{V}{V} \quad \beta = \frac{1}{13} \frac{V}{V} \Rightarrow \boxed{GBP = A_0 \cdot \beta \cdot f_t = 13 \cdot \frac{1}{13} \cdot 4 \text{ MHz} = 4 \text{ MHz}} = f_B$$

(c)

Left in the air



$$v_O = 5v_I \Rightarrow \boxed{A_0 = 5 \frac{V}{V}}$$

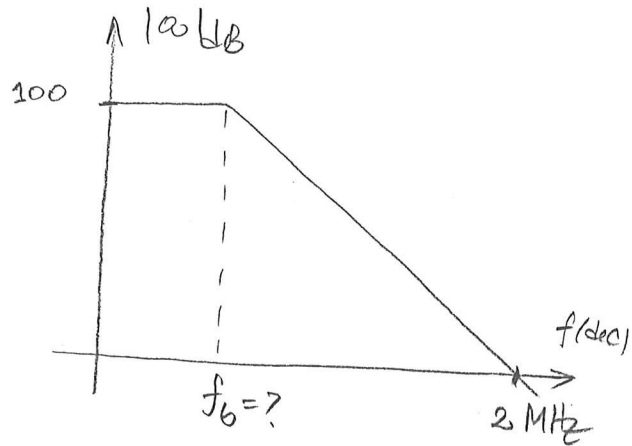
$$A_0 = 5 \frac{V}{V} \quad \beta = \frac{1}{5} \Rightarrow \boxed{GBP = 4 \text{ MHz}}$$

(Exam 1; Spring '11)

PROBLEM 5

The open-loop frequency characteristic of an OpAmp, using the dominant pole model, is shown in the figure below. The OpAmp has $r_d = 1\text{M}\Omega$ and $r_o = 100\Omega$.

- Find the value of f_b for the OpAmp.
- If the OpAmp is used to build a non-inverting amplifier with a gain of 10, calculate the bandwidth of the amplifier. Calculate the input impedance for an input frequency of 500kHz.



(a) $a_0 = 10^{100/20} = 10^5$

$f_t = 2\text{MHz} = f_b \cdot a_0 \Rightarrow$

$\Rightarrow f_b = \frac{f_t}{a_0} = 20\text{Hz}$

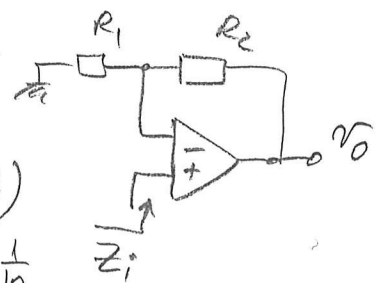
(b) $A_0 = 10$, $f_t = A_0 \cdot f_B$ $\Rightarrow f_B = \frac{f_t}{A_0} = 200\text{kHz}$.
(Bandwidth!)

The input impedance of a "series input" negative feedback configuration is:

$Z_i = R_{id} \frac{1 + j f/f_B}{1 + j f/f_0}$, where: $R_{id} = r_d(1 + a_0\beta)$

and $\beta = \frac{R_1}{R_1 + R_2} = \frac{1}{10}$

$\Rightarrow R_{id} = 1\text{M} \left(1 + 10^5 \cdot \frac{1}{10}\right) \approx 10^6 \cdot 1\text{M}\Omega = 10\text{G}\Omega$



$f = 500\text{kHz} \Rightarrow |Z_i| = R_{id} \cdot \left| \frac{1 + j f/f_B}{1 + j f/f_0} \right|$

$= 10\text{G}\Omega \cdot \frac{\sqrt{1 + \left(\frac{500\text{k}}{20\text{k}}\right)^2}}{\sqrt{1 + \left(\frac{500\text{k}}{20}\right)^2}}$

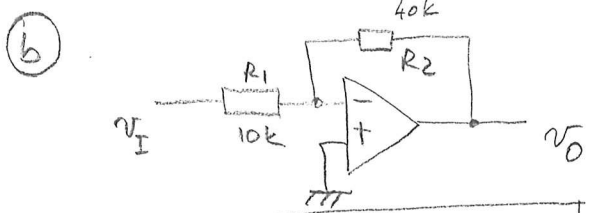
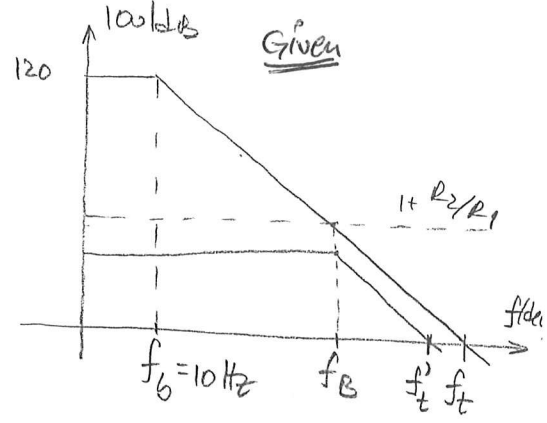
EXAM 2- ECE 723 - SPRING 2009
Open Book, Closed Notes, 50 minutes

PROBLEM 1

The open-loop frequency characteristic of an OpAmp, using the dominant pole model, is shown in the figure below. The OpAmp has $r_d = 1M\Omega$ and $r_o = 100\Omega$, $C_c = 30pF$.

- Find the value of f_t for the OpAmp.
- If the OpAmp is used to build an **inverting** amplifier with $R_1=10k$ and $R_2=40k$, sketch its frequency response and calculate:
 - the 3dB bandwidth (that is f_B)
 - the unity gain bandwidth of the amplifier (that is f'_t)
- If the OpAmp is used to build an **inverting** amplifier with a gain of 10, derive the formula calculate the frequency f_x at which the input impedance is $1.2R_1$. Assume that such frequency $f_b \ll f_x \ll f_t$.

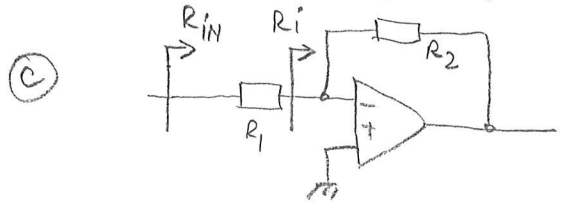
(a) $120 \text{ dB} = 20 \cdot \log a_0 \Leftrightarrow 120 = 20 \cdot \log a_0$
 $\Rightarrow a_0 = 10^6$
 $f_t = a_0 \cdot f_b = 10^6 \cdot 10 \text{ Hz} = 10 \text{ MHz}$



$A(jf) = A_0 \cdot \frac{1}{1 + j f/f_B}$ Eq. 6.15.a

$|A_d| = \left| -\frac{R_2}{R_1} \right| = 4$ $f_B \approx \beta \cdot f_t = \frac{R_1}{R_1 + R_2} \cdot f_t = \frac{1}{5} \cdot 10 \text{ MHz} = 2 \text{ MHz}$ $f_B = 2 \text{ MHz}$

Use Eq. 6.16 $\Rightarrow f'_t = (1 - \beta) f_t = \frac{4}{5} \cdot 10 \text{ MHz} = 8 \text{ MHz}$ $f'_t = 8 \text{ MHz}$



Use Eq. 6.23 $\Rightarrow R_{IN} = R_1 + R_i = R_1 + \frac{R_2}{1 + a_0} \cdot \frac{1 + j f/f_b}{1 + j f/f_t} \Rightarrow$

$\Rightarrow 1.2 \cdot R_1 = \left| R_1 + \frac{R_2}{1 + a_0} \cdot \frac{1 + j f_x/f_b}{1 + j f_x/f_t} \right| \approx \left| R_1 + \frac{R_2}{1 + a_0} \cdot (j \cdot \frac{f_x}{f_b}) \right| \Rightarrow$

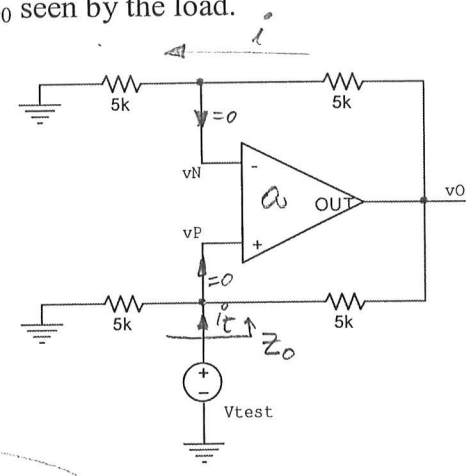
$\Rightarrow 1.2^2 = 1 + \left(\frac{R_2}{R_1} \cdot \frac{1}{1 + a_0} \right)^2 \cdot \left(\frac{f_x}{f_b} \right)^2 \Rightarrow f_x = \dots$

(This is similar to last problem of Exam #11)

PROBLEM 2 (Exam 2; Spring 11)

P. 6.24 ?

In the given circuit, all resistances are equal to $5k\Omega$. The Op Amp has $a_0 = 10^5$ V/V, $f_t = 10$ MHz, $r_d = \infty$, $r_o = 0$. Sketch the magnitude plot of the impedance Z_0 seen by the load.



$$v_{test} = v_P$$

$$\left\{ \begin{aligned} v_O &= a_0 \cdot (v_P - v_N) = a_0 \cdot (v_{test} - v_N) \\ v_O &= 2 \cdot v_N \end{aligned} \right\} \Rightarrow$$

$$\Rightarrow v_N = \frac{a_0}{2 + a_0} \cdot v_{test}$$

$$i_t = \frac{v_{test}}{R} + \frac{v_{test} - v_O}{R} = \frac{v_{test} + v_{test} - 2v_N}{R} = \frac{2}{(1 + \frac{a_0}{2})R} \cdot v_{test} \Rightarrow$$

$$\Rightarrow Z_0 = \frac{v_{test}}{i_t} = \frac{R}{2} \left(1 + \frac{a_0}{2} \right)$$

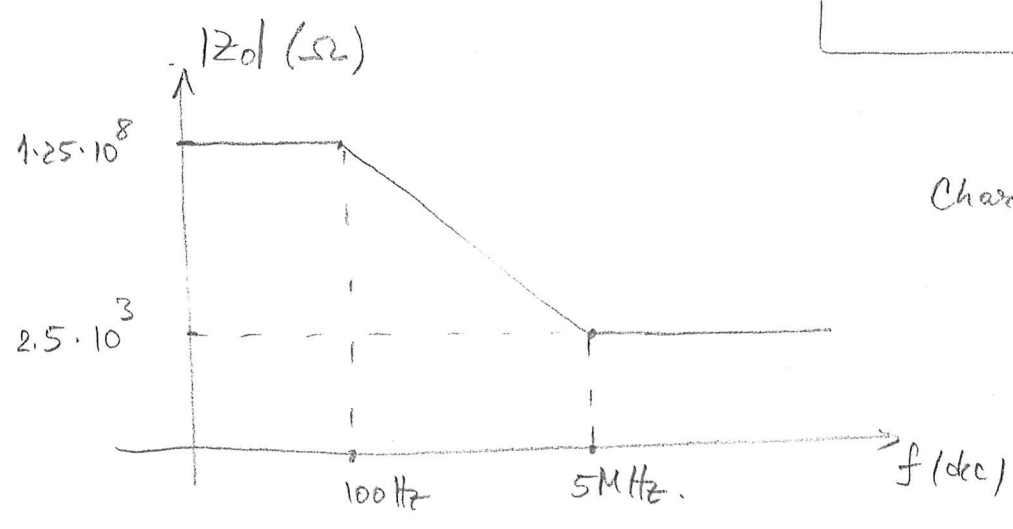
We know:

$$a_0 = \frac{a_0}{1 + j \frac{f}{f_b}}$$

where $f_b = f_t / a_0 = 100\text{Hz}$

$$\Rightarrow Z_0 = (2.5k\Omega) \cdot \left(1 + \frac{10^5/2}{1 + j \frac{f}{100\text{Hz}}} \right)$$

$$\approx 1.25 \cdot 10^8 \cdot \frac{1 + j \frac{f}{5\text{MHz}}}{1 + j \frac{f}{100\text{Hz}}} = Z_0$$



Character capacitive!

(This is similar to a problem from HW #5!)