

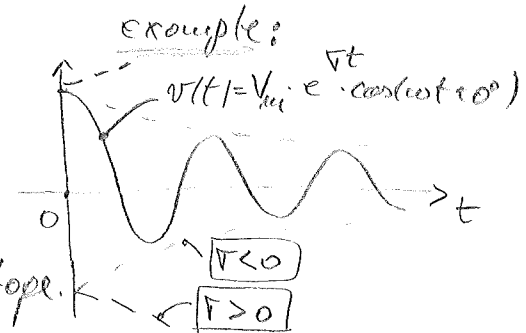
Chapter 14

Complex frequency and the Laplace transform

1. Complex frequency

- Consider an exponentially damped sinusoidal function:

$$(1) \quad v(t) = V_{\text{re}} \cdot e^{\gamma t} \cdot \cos(\omega t + \theta)$$



let $\begin{cases} \gamma = 0 \\ \omega = 0 \end{cases}$

$$(2) \quad v(t) = V_{\text{re}} \cos \theta = V_0 \quad \text{a dc voltage.}$$

$\gamma = 0$

$$(3) \quad v(t) = V_{\text{re}} \cos(\omega t + \theta) \quad \text{an ac voltage.}$$

all special cases of

$\omega = 0$

$$(4) \quad v(t) = V_{\text{re}} \cdot e^{\gamma t} \cdot \cos \theta = V_0 \cdot e^{\gamma t} \quad \text{the exponential voltage.}$$

- Remember: the complex representation of a sinusoidal function with phase-angle zero:

$$(5) \quad v(t) = V_0 \cdot e^{j\omega t}$$

- Similarity between (4) & (5) is emphasized by describing γ as "frequency". More precisely as the "real part" of the complex frequency.

↑
defined next page.

General form

- A "complex exponential signal" is a signal of the form:

$$x(t) = X \cdot e^{st} \quad (6)$$

Note: the 1st time we mix phasors and time variable!

- X, s are "time-independent complex parameters" expressed in polar and rectangular coordinates as:

Note, it's a phasor!

$$X = X_{m} \cdot e^{j\theta}$$

phase angle (radians or degrees)

(7)

magnitude of $x(t)$ may be [V or A]

$$s = \sigma + j\omega$$

real part of $s \triangleq$ Neper frequency in [nepers/s]

imaginary part of s ; called radian frequency [radians/s]

Note: use $x(t)$ because in general can be a voltage or current!

\triangleq complex frequency

$$\Rightarrow x(t) = X_{m} \cdot e^{j\theta} \cdot e^{(\sigma + j\omega)t} = X_{m} \cdot e^{\sigma t} \cdot e^{j(\omega t + \theta)} \quad (6')$$

- Using Euler's formula, we have the following alternate form for a complex exponential signal:

Note: this is a complex function!

$$x(t) = X_{m} \cdot e^{\sigma t} \cdot \left[\cos(\omega t + \theta) + j \sin(\omega t + \theta) \right] \quad (8)$$

Because it's complex, it cannot be duplicated in laboratory, where signals are real. However, mathematicians showed that real-world signals can be expressed as linear combinations of complex exponential signals. Knowing the response of a circuit to individual complex exponential signals, we can apply the "superposition principle".

Examples:

- Generally: $x(t)$ can be written as:
a real:

$$x(t) = \frac{1}{2} (x + x^*) = \text{Re} \{ x \}$$

real function we are interested to use in circuits. Not bold character.

Complex variables/functions.

Complex exponential signal from equation (8)

its conjugate.

should be written with bold characters to show that they are complex.

- Actual voltage signal can be written as:

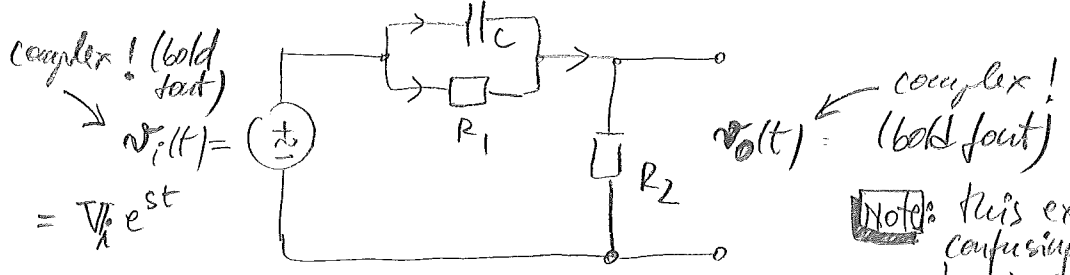
$$v(t) = \text{Re} \left\{ \underbrace{V_m e^{st} \cdot e^{j(\omega t + \theta)}}_{\substack{\uparrow \\ \text{complex} \\ \text{exponential} \\ \text{signal}}} \right\} = V_m e^{st} \cdot \cos(\omega t + \theta)$$

a exponentially varying sinusoidal

Remember: how we developed the frequency-domain descriptions of ac circuits, when we introduced the complex forcing function $[\cos \omega t = \text{Re}(e^{j\omega t})]$ and we worked with complex source $V_m e^{j\omega t}$; response was complex whose real part corresponded to the real part of the complex source! **NOTE:** Moreover, we have introduced phasors, which "carry" info on amplitude/magnitude and phase; ω was implicit/understood.

Obs: Here, we simply extended the notations to include the damped sinusoidal forcing function at a complex frequency. The difference is that here we replaced $j\omega$ with s .

Example 1 develop a quick feel for how to handle complex exponential signals: (4)



Note: This example may be a bit confusing due to $v(t)$ time domain and $v(t)$ complex!?

$v_o(t) = V_o \cdot e^{st}$ and find relationship between $v_i(t)$ and $v_o(t)$

Solution:

KCL: $C \frac{d(v_i - v_o)}{dt} + \frac{v_i - v_o}{R_1} = \frac{v_o}{R_2}$

↑
here these are complex variables

$$\left. \begin{aligned} R_1 R_2 C \frac{dv_o}{dt} + (R_1 + R_2) v_o &= R_1 R_2 C \frac{dv_i}{dt} + R_2 v_i \\ v_i &= V_i e^{st} \\ v_o &= V_o e^{st} \end{aligned} \right\} \Rightarrow$$

$\Rightarrow R_1 R_2 C s V_o e^{st} + (R_1 + R_2) V_o e^{st} = R_1 R_2 C s V_i e^{st} + R_2 V_i e^{st}$

$\Rightarrow v_o(t) = \frac{R_1 R_2 C s + R_2}{R_1 R_2 C s + (R_1 + R_2)} \cdot v_i(t)$

Example 2 In previous example for $R_1 = 1 \Omega$, $R_2 = 2 \Omega$, $C = 1 F$, given:

$v_i(t) = 10 \cdot e^{-0.5t} \cdot \cos(1.5t + 30^\circ) V$ find: $v_o(t) = ?$

$v_i(t) = \text{Re} \{ v_i(t) \} = \text{Re} \{ 10 \angle 30^\circ \cdot e^{(-0.5 + j1.5)t} \}$

↑
real time-domain

↑
complex exponential signal should be understood from context.

The complex response $v_0(t)$ to the complex signal $v_i(t)$ is: (5)

$$v_0(t) = \frac{2(-0.5 + j1.5) + 2}{2(-0.5 + j1.5) + 3} \cdot 10 \angle 30^\circ \cdot e^{(-0.5 + j1.5)t}$$
$$= 8.971 \angle 45.26^\circ \cdot e^{(-0.5 + j1.5)t} \quad \text{V}$$

The real response then $v_0(t)$ to the real signal $v_i(t)$ is:

that is
time domain

$$v_0(t) = \text{Re} \{ v_0(t) \}$$

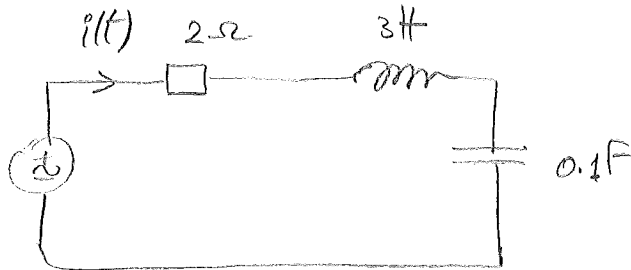
or:
$$v_0(t) = 8.971 \cdot e^{-0.5t} \cdot \cos(1.5t + 45.26^\circ) \quad \text{V}$$

which is a damped sinusoid with the same damping factor σ and frequency ω as the applied signal, differing only in amplitude and phase!

Example 3Given:

$$v(t) = 60 \cdot e^{-2t} \cos(4t + 10^\circ)$$

\uparrow \uparrow \uparrow \uparrow
 V_m \downarrow ω θ

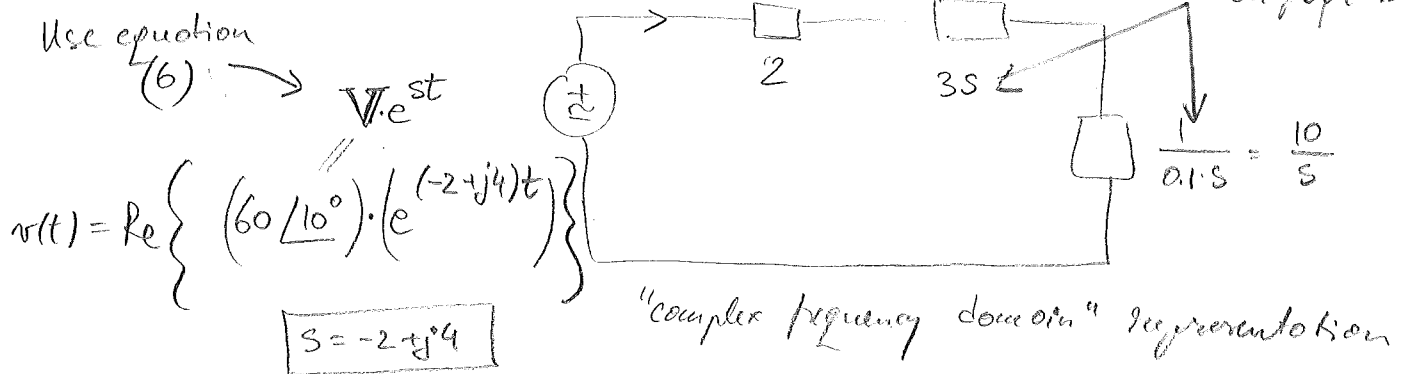


Find: $i(t) = I_m e^{-2t} \cos(4t + \phi)$

\uparrow \uparrow
 I_m ϕ

We expect: $I e^{st}$

based on discussion on page 3!



- Apply KVL:

$$60 \angle 10^\circ e^{st} = 2 I e^{st} + 3S I e^{st} + \frac{10}{S} I e^{st}$$

$$60 \angle 10^\circ = 2 I + 3S I + \frac{10}{S} I$$

$$I = \frac{60 \angle 10^\circ}{2 + 3S + 10/S} \Bigg|_{s = -2 + j4} \Rightarrow I = \frac{60 \angle 10^\circ}{2 + 3(-2 + j4) + 10/(-2 + j4)} = 5.37 \angle -106.6^\circ$$

Therefore: $I_m = 5.37 \text{ A}$, $\phi = -106.6^\circ$ and the time-domain real forced response is:

$$i(t) = 5.37 \cdot e^{-2t} \cdot \cos(4t - 106.6^\circ) \text{ A}$$