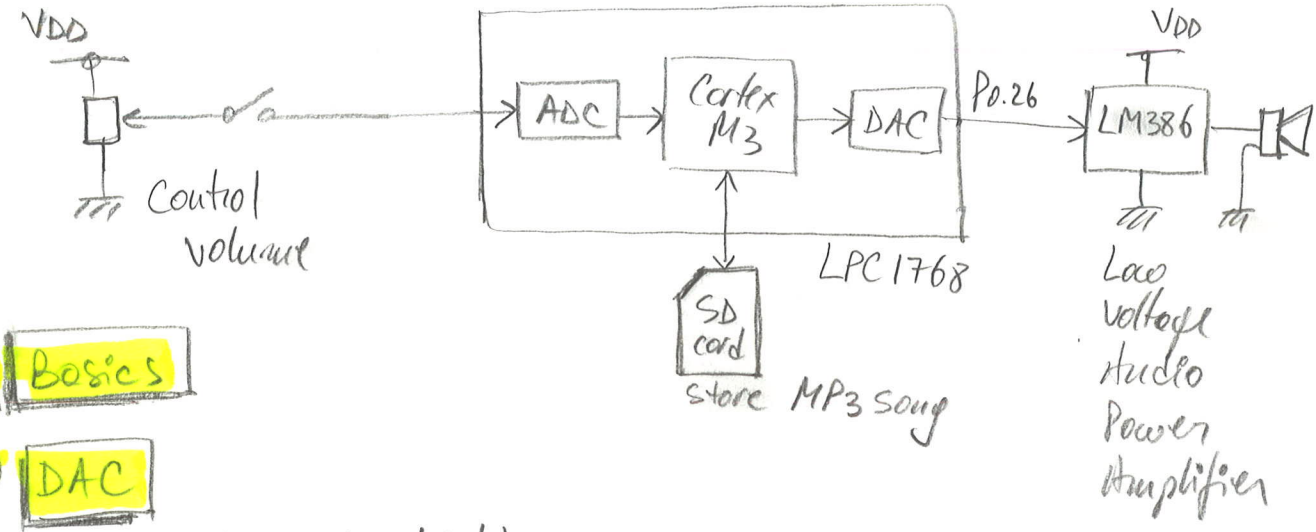
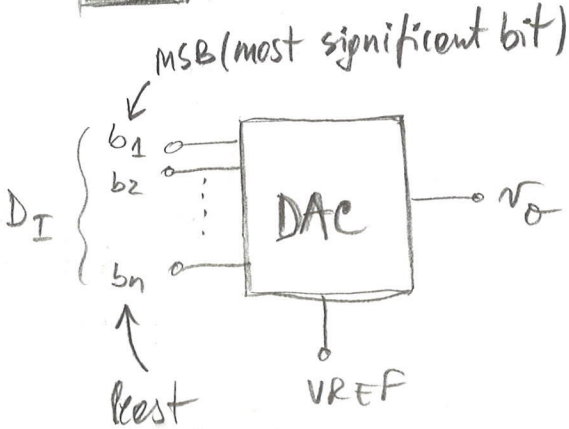
Introduction to (1st talk about IRQ vs. FIQ)

Digital to Analog Converters DAC and Analog to Digital Converters ADC

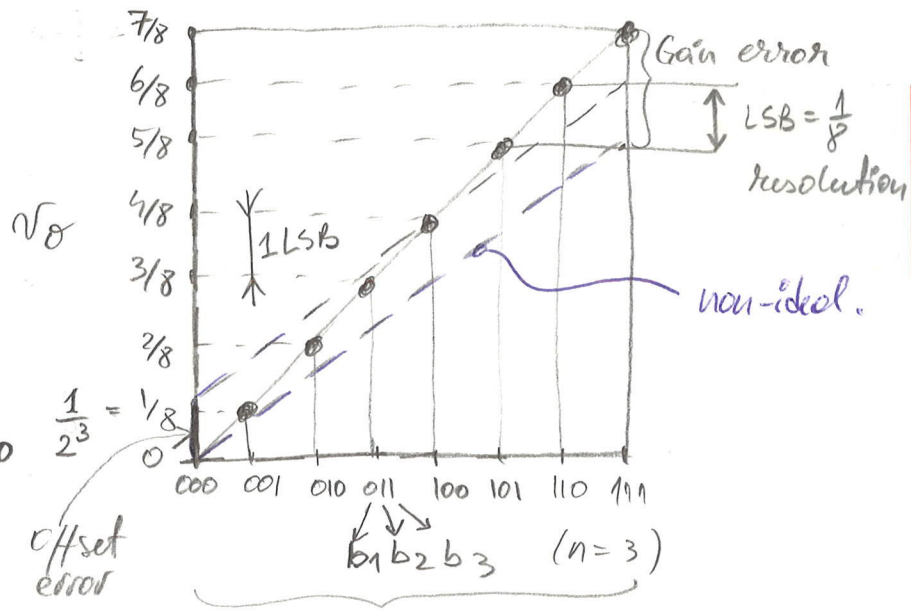


A Basics

1 DAC



Example: $(b_1 b_2 b_3 b_4)_2 = (0111)_2 = 7_{10}$



2^n points $\Rightarrow$ 2^n output values

(1) $\boxed{v_O = k V_{REF} \cdot D_I}$

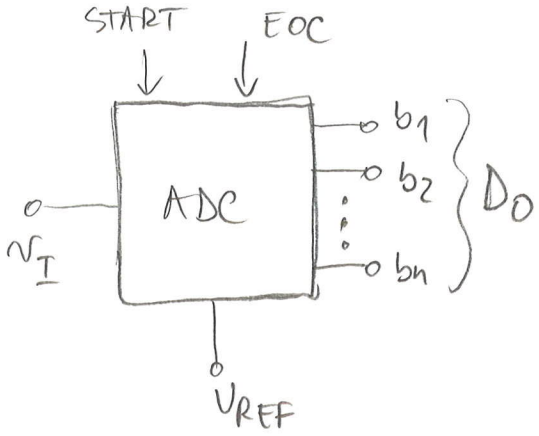
$= k V_{REF} \left(b_1 \cdot \frac{1}{2} + b_2 \frac{1}{2^2} + \dots + b_n \frac{1}{2^n} \right)$ (MSB, LSB)

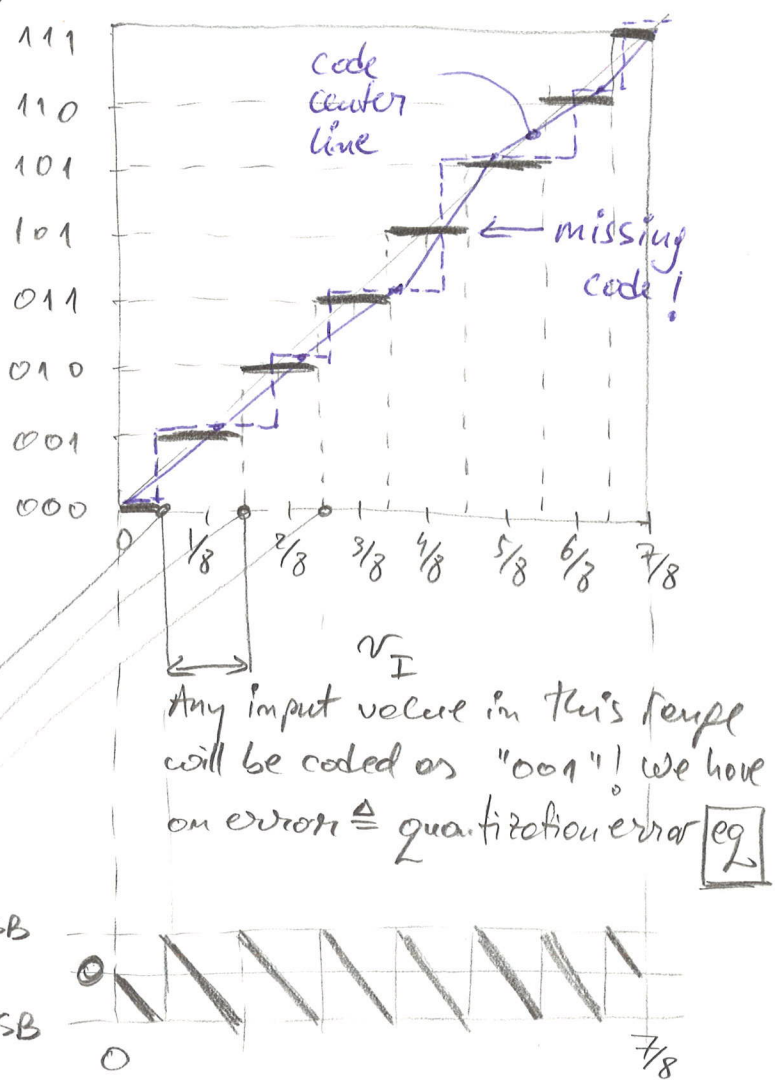
$= k \cdot V_{REF} \cdot \sum_{i=1}^{n} \frac{b_i}{2^i}$

Definitions:
- FSR ≡ full scale range $V_{FSR} = k V_{REF}$
- FSV ≡ full scale value $V_{FSV} = (1 - 2^{-n}) V_{FSR}$
- LSB ≡ Resolution $\frac{V_{FSR}}{2^n}$

- This is called a **multiplying DAC** because v_O is obtained by multiplying $V_{REF} \times D_I$: $D_I = b_1 \cdot 2^{-1} + b_2 \cdot 2^{-2} + \dots + b_n \cdot 2^{-n}$

(2) **ADC** provides the inverse function of DAC!





$$D_O = b_1 \frac{1}{2} + b_2 \frac{1}{2^2} + \dots + b_n \frac{1}{2^n}$$

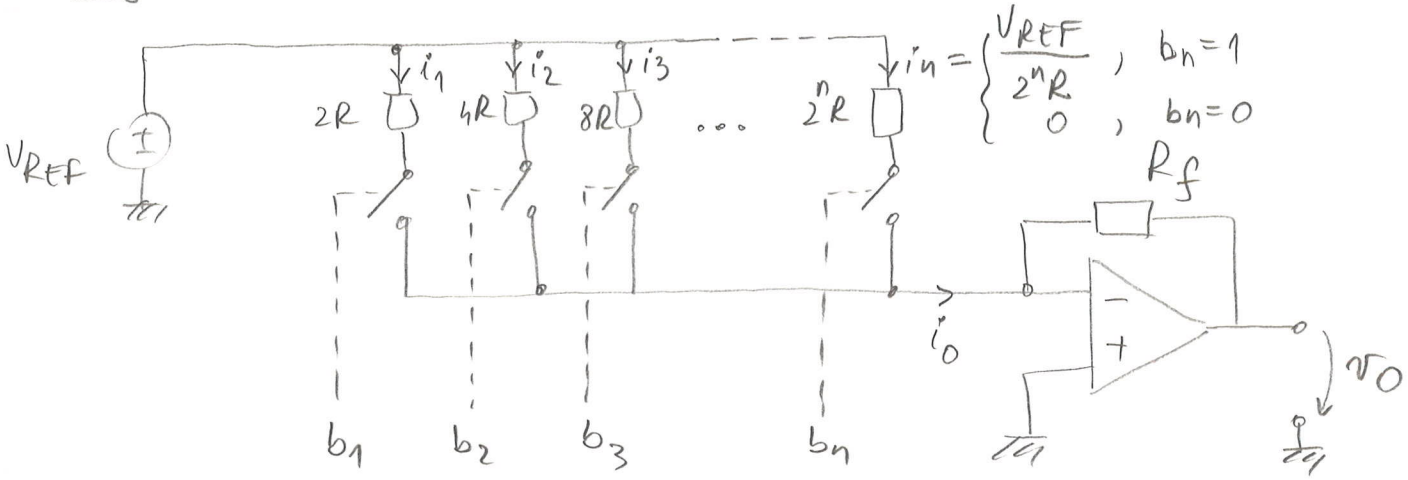
$$= \frac{V_I}{k \cdot V_{REF}} = \frac{V_I}{V_{FSR}}$$

ideally these transitions take place at odd numbers of $\frac{1}{2}$LSB; if not, then we have nonlinearities (errors)

- Conversion time ≙ time for ADC to complete the transition!

B DA Techniques

(a) Weighted resistor DAC



$$\frac{v_0}{R_f} = -i_0$$

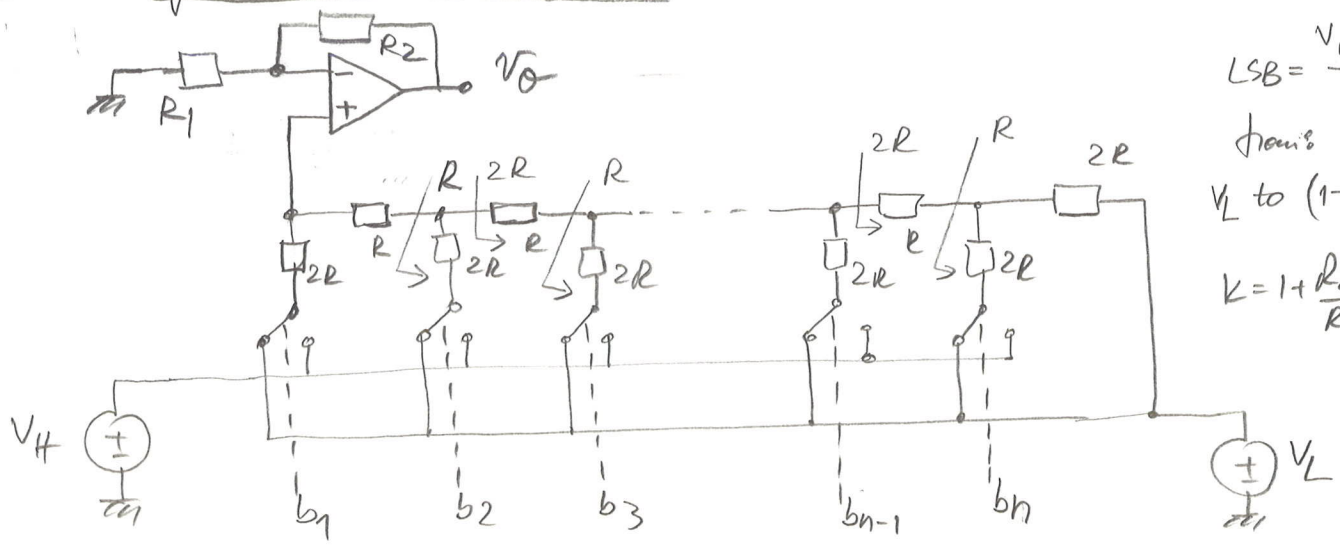
$$v_0 = -\frac{R_f}{R} \cdot V_{REF}\left(b_1 \frac{1}{2} + b_2 \frac{1}{4} + b_3 \frac{1}{2^3} + \ldots + b_n \frac{1}{2^n}\right)$$

Drawbacks – monotonic switch resistance
– wide spread of resistor values (exponential increase!)

(b) Weighted-Capacitor DAC

(c) Potentiometric DAC

(d) Voltage-mode R-2R ladder



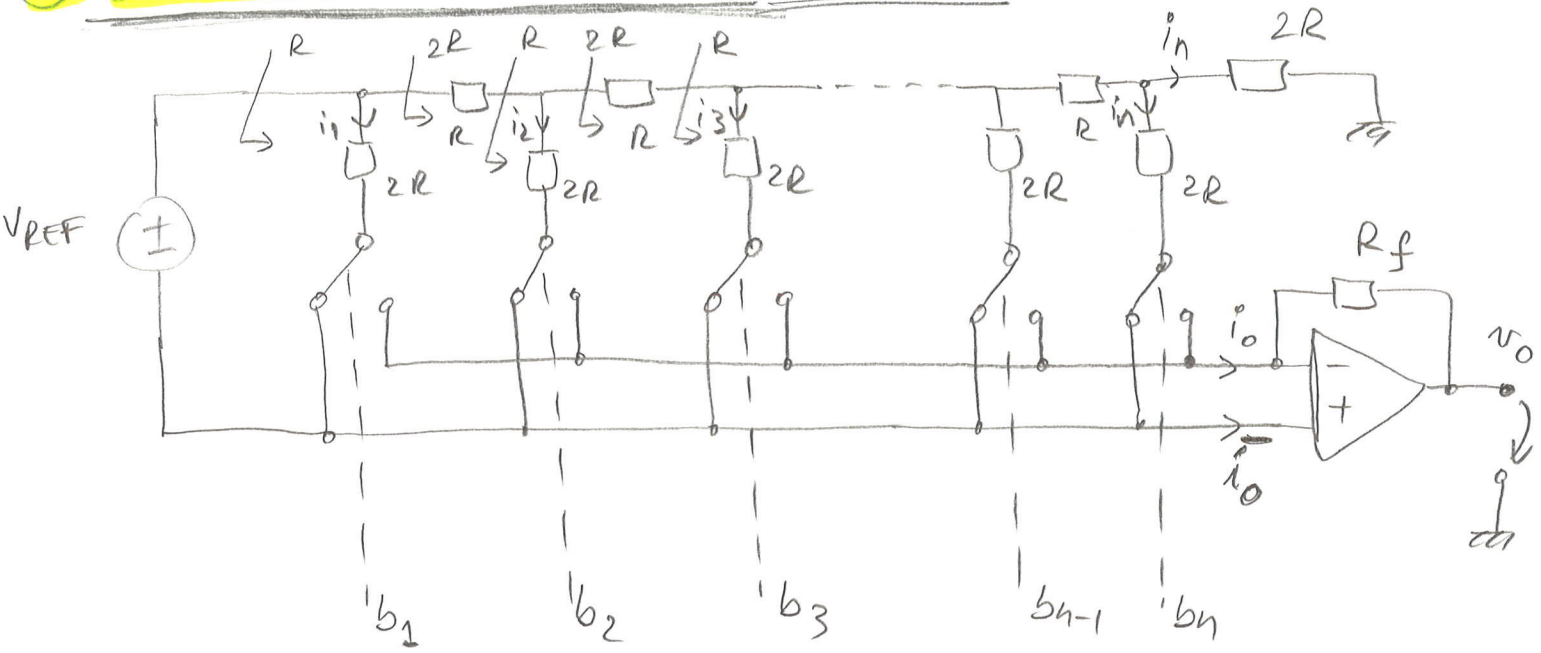
$$LSB = \frac{V_H - V_L}{2^n}$$

from V_L to $\left(1 - \frac{1}{2^n}\right)(V_H - V_L)$

$$K = 1 + \frac{R_2}{R_1}$$

Advantage: we can interpolate between only V_L, V_H !

② Current-mode R-2R ladder based DA



$$\begin{cases} i_1 = \dfrac{V_{REF}}{2R} = \dfrac{V_{REF}}{R}\cdot\dfrac{1}{2} \\ i_2 = \dfrac{V_{REF}}{4R} = \dfrac{V_{REF}}{R}\cdot\dfrac{1}{2^2} \\ \dots \\ i_n = \dfrac{V_{REF}}{R}\cdot\dfrac{1}{2^n} \end{cases}$$

$$\frac{v_o}{R_f} = -i_o$$

$$v_o = -R_f \cdot i_o$$

OBS:- Note that $i_o + \bar{i}_o = (1-2^{-n})\cdot\dfrac{V_{REF}}{R}$

regardless of $D_I = b_1 b_2 b_3 \dots b_n$!

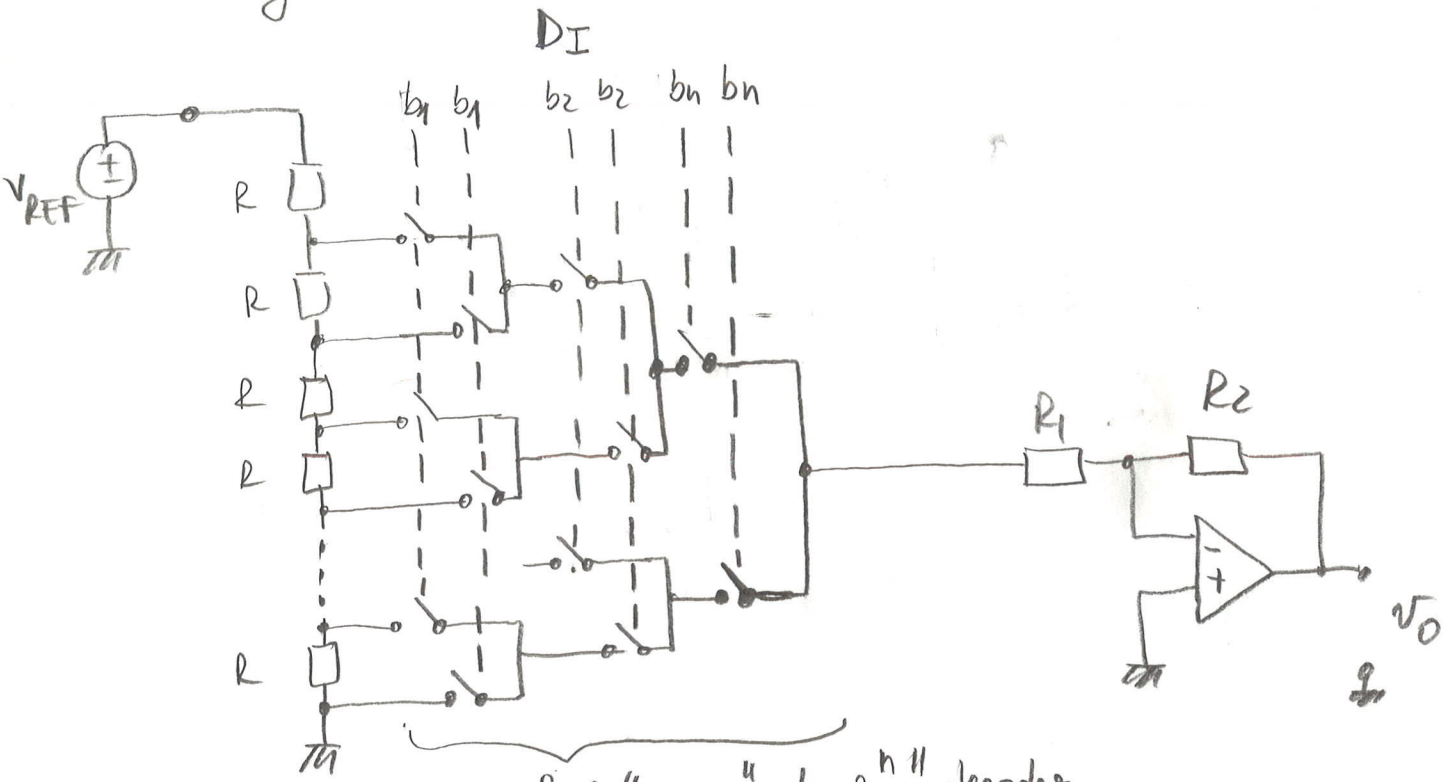
- $\bar{i}_o$ is said to be complementary to i_o

$$v_o = -\frac{R_f}{R} V_{REF}\cdot\left(b_1\frac{1}{2} + b_2\frac{1}{2^2} + \dots + b_n\frac{1}{2^n}\right)$$

$$\triangleq K = -\frac{R_f}{R}$$

$$\boxed{v_o = K V_{REF}\left(b_1\frac{1}{2} + b_2\frac{1}{2^2} + \dots + b_n\frac{1}{2^n}\right)}$$

(f) R-String architecture (used inside LPC1768)



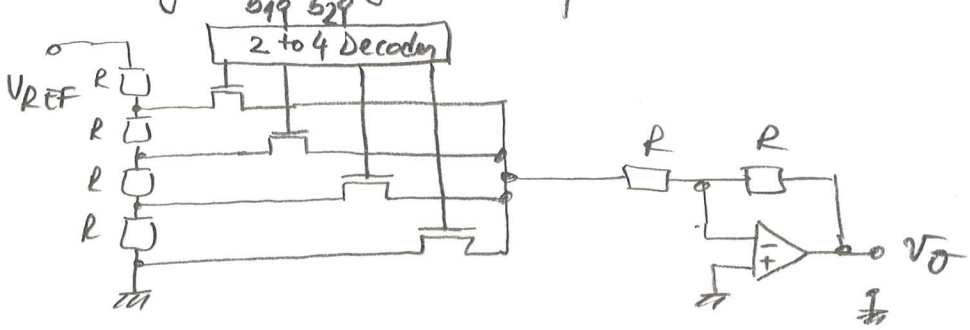
≡ Basically a "n to 2^n" decoder.

Advantages: - simple (needs only identical resistors)
- fast for < 8-10 bits
- monotonic (inherently)
- compatible w/ purely digital technologies
- low power

Disadvantages:

- 2^n resistors and 2^n switches for n bits
  => high element count, high area!
- high settling time for n > 10

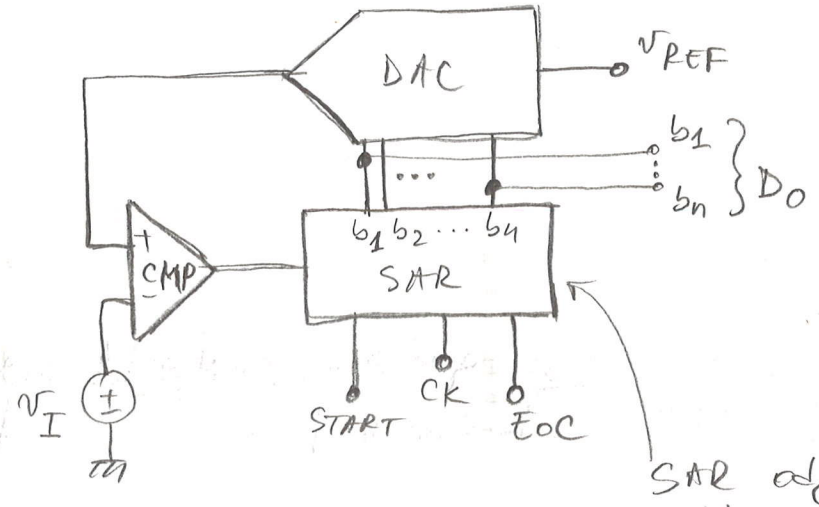
Example:



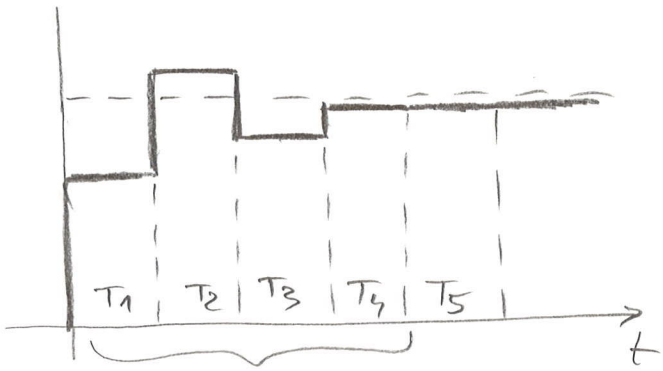
C A/D techniques

a) DAC based AD conversion:
"successive-approximation converters" ← uses SAR (successive approximation register)
b) charge-redistribution converters
c) flash converters
d) subranging converters
e) integrating-type converters.

a) DAC based A-D Conversion (used by LPC 1768)



Example:



SAR adjusts DAC's input code until DAC's output comes within ½ LSB of input voltage V_I!

c) Flash converters